Design and Simulation of MEMS based Inertial Electromagnetic

Harvester



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A thesis submitted in partial fulfillment of the requirements for the degree of MS Mechatronics Engineering

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I certify that this research work titled "*Design and Simulation of MEMS based Inertial Electromagnetic Harvester*" is my own work. The work has not been presented elsewhere for assessment. The material that has been used from other sources it has been properly acknowledged / referred.

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Language Correctness Certificate

This thesis has been read by an English expert and is free of typing, syntax, semantic, grammatical and spelling mistakes. Thesis is also according to the format given by the university.

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Abstract

Due to the advent in low power wireless electronics, devices are becoming compact and power requirements have reduced to the levels of microwatts. So intensive research is being carried on to scavenge energy from ambient sources found in environment to power portable electronics and ultra-low power wireless sensors by external sources. This thesis investigates the development of one of the technique of harvesting electrical energy from vibrations based on the Faraday's law of electromagnetic induction. Focus was to design a harvester completely compatible with MEMS fabrication techniques, can work on multiple frequencies and to take advantage of magnetic flux distribution of an array of permanent magnets in the specific direction (also called Active side) by using Halbach Array of magnets. The micro fabricated coils connected in series with each other are placed towards an Active side. Finally, the proposed design resulted in a maximum output voltage of 18mV with a maximum power of 0.53nW at a resistive load of 98k Ω and frequency of 40.7Hz.

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List of Parameters

Name	Expression	Value	Description
Lmx	8[mm]	0.008 m	Total x length of magnets
Lmy	5[mm]	0.005 m	Total y length of magnets
f	48.9[Hz]	48.9 Hz	Frequency
Т	1/f	0.02045 s	Period
Br	1.4[T]	1.4 T	Remanent flux density in PM
dis	30[um]	3E-5 m	Distance (gap) between magnets and coils
Lcx	Lmx/4	0.002 m	x-length of coil
Lcy	Lmy	0.005 m	y-length of coil
xc	Lmx/2	0.004 m	x-coordinate of coil border (half length in x
			direction)
yc	Lmy/2	0.0025 m	y-coordinate of coil border (half length in y
			direction)
Nc	52	52	Number of turns in each real coil
Ntotal	Nc*8	416	Total number of turns
c	Lmy/16	3.125E-4 m	Width of each magnet
fac	2	2	Factor of nominal frequency for filter
omegan	2[rad]*pi*f	307.25 rad/s	Nominal frequency
omegac	fac*omegan	614.5 rad/s	Cut-off angular frequency (for filter)
RC	98[kohm]	98000 Ω	Resistance of coils
LC	0.00353[H]	0.00353 H	Inductance of coils
RF	RC	98000 Ω	Resistance in filter
CF	1/(RF*omegac)	1.6606E-8 F	Capacitance in filter
RL	RC	98000 Ω	Resistance of load
CL	1/(LC*omegan^2)	0.0030009 F	Capacitance of load

Chapter 1. INTRODUCTION

Electrical energy has become the key resource in development of any nation. Since the dawn of industrial age humans are constantly searching for the reliable energy sources. As the time passed relying on electrical energy has become more and crucial as well as a constant threat to the earth's ecology. Therefore, a race of renewable power sources has been started. Many natural resources have been employed to extract energy like solar, wind, water and thermal. But these energy providers are bulk in size, expensive and need constant maintenance. So to power up low consumption devices like low power electronic circuits in remote areas such sensor networks, different energy harvesting techniques were proposed so far.

1.1 Energy Harvesting

The term, Energy harvesting was used in literature to describe the process of photosynthesis in plants due to sunlight. Now this term readily used for extracting energy from ambient resources like solar radiation, motion of natural of artificial objects and thermal processes in nature [1]. Energy harvesting is basically a process of harnessing environmental sources to generate electricity on micro level. This energy can be efficiently utilize in sensors and other low power circuits which require the power of few micro watts [2]. Energy harvester have many uses from military applications to wearable and medical devices.

1.1.1 Background

The basis of energy harvesting is directly derived from Faraday's Law of electromagnetic induction in 1831 [3]. This law of electromagnetic induction provides the possibility to convert mechanical energy in electrical energy. Energy harvesting system works on the principle of change of flux in the coil of a harvester when it oscillates due to the ambient motion from environment. The induced voltage and energy will then be supplied to attached load. Energy harvesters find their applications in many devices and can be used on several vibrating sources. Electro Magnetic Micro Harvesters will be discussed in detail in later sections.

1.2 Micro Energy Harvesters

Micro Energy harvesters are considered as one of the best wireless power suppliers. They have the same importance as that of wireless communication because of the several reasons like portability, no need of replacement of batteries, reduced installation cost and continuous supply of power to remote locations. Previously in such applications batteries were used as power supplier but due to their several disadvantages, efforts are being made to replace batteries by micro energy harvesters. Also an increase in mobile electronic devices is giving rise to the need of wireless sources of power Due to this reason, mechanical vibrations are getting considerable amount of interest of researchers as a source of power supply. Since environment supply an ample amount of vibrations, these vibrations can be converted into useful amount of electrical energy by using self-contained and maintenance free energy harvesters. A common example of this phenomenon is the utilization of continuous vibrations during the operation of a large range of industrial machinery such as electrical motors, compressors, turbines etc. Micro Level vibration energy harvesters can be installed on these machines to power up the sensors which can monitor condition and health of the machine. Another common example of this type of usage is in the automotive industry for automatic monitoring of tire pressure[4] [5].

1.2.1 Types of Micro Energy Harvester

In this thesis, vibration-based energy harvesters are of major consideration. These type of micro generator consist of inertial mass attached to the spring. This inertial mass vibrates in to and fro motion relative to frame when subjected to vibration. This motion of inertial mass crosses the magnetic flux and generate electricity as a result. Based on the above discussion, micro energy harvesters can broadly categorize in three types;

- 1) Piezoelectric Energy Harvesters
- 2) Electrostatic Energy Harvesters
- 3) Electromagnetic Energy Harvesters

1.2.1.1 Piezoelectric Energy Harvesters

The piezoelectric energy harvesters are of much interest from the last two decades. Piezoelectric energy harvesters work on the principle of strain generation in piezoelectric materials such as PZT or PVDF. In principle, when vibration from environment deform the piezoelectric material, a

potential difference will be generated across the terminals. The produced potential difference forms a damping force which is proportional to relative velocity in the structure made of piezoelectric material. This property of voltage generation when subjected to strain, is due to sudden charging within the crystals, polycrystalline ceramics and piezoelectric polymers.

The most basic design of piezoelectric harvester is like cantilever beam mentioned in literature [6]. It is manufactured by depositing thin layer of piezoelectric material on micro fabricated cantilever beam with lump mass at the free end of the beam. The lumped mass controls the frequency of the beam when subjected to vibration. Figure 1.1 depicts how piezoelectric harvester works in principle [7].



Figure 1.1: Conceptual design of Piezoelectric Energy harvester [7].

There are two type of operations that can be performed on a piezoelectric harvester. From Figure 1.1 it is clear that force can be either applied along with axis 3 and axis 1. The generated output voltage will obtain across the face perpendicular to axis 3 and a simple resistive load can be applied to estimate the amount of power or optionally a storage capacitor can also be used in cases where data need to be received and transmitted for few moments and this cycle continues [8]. Also composite beam piezoelectric harvesters are most commonly used and are of much interest due to their wide no of applications and effectiveness [9].

However piezoelectric generators are very difficult to manufacture using standard MEMS fabrication methods. These devices perform on very low vibration and they are vulnerable to mechanical defects like creep and fatigue over accidental or repetitive/cyclic loading and voltage breakdown at high frequencies [10]. Also piezoelectric harvesters may suffer from charge leakage at low operating frequencies [11] along with the issues of high internal resistance, low output current and low power density make them less suitable. [12][13][14].

1.2.1.2 Electrostatic Energy Harvesters

An electrostatic energy harvester works on the principle of gap variation between two conductive plates resulting in the change of capacitance. The gap is filled with some dielectric, in most cases air. The capacitance between two plates changes with when the conductive plates moves relative to each other due to vibration. This results in increase in stored energy in the harvester [15]. Also electrostatic transducers are much suitable at low frequencies and their integration with microelectronics is also feasible.

In electrostatic energy harvesters, configuration of capacitance of the system will decide to be a voltage source or a current source. In either of the case, one parameter either voltage or current is kept as constant. Keeping the charge constant across the capacitor, the device will act as voltage source. In this condition an additional charge source is required to keep the charge constant. However, if the potential difference is kept constant then the device will act as charge source. In this case, an additional voltage source will be required to keep the voltage constant. There are three mechanical configurations to use the capacitance as energy.

In-Plane overlap conversion

In-Plane gap closing conversation

Out-of-plane gap closing conversation

All these conversations can be seen in figure below



Figure 1.2: (a) In-Plane Overlap. (b) In-Plane Gap Closing. (c) Out-of-Plane Gap Closing [16].

But electrostatic harvesters are effective at microscale when they are employed for actuation purposes but their results are poor when used as generators [17] Another issue of electrostatic harvesters is their preliminary requirement of an additional source of voltage to pre-charge [15], however in many harvesters this problem is over-come by the usage of electrets [18]. Also electrostatic harvesters are much feasible for the use at low frequencies and low stiffness applications due to their low losses [10] [19].

1.2.1.3 Electromagnetic Energy Harvesters

Electromagnetic Micro Energy Harvesters (EMEHs) works on the principle of Faraday's Law of Electromagnetic Induction. According to Faraday's Law when a conductor moves in a magnetic flux, a potential difference will be created on the ends of inductor.

EMEHs produce electricity due to the motion of coil in the magnetic field of a permanent magnet. Normally, coil could be a wounded wire over an insulator or just planer wires over an insulator. When EMEH is supplied by some vibration from an ambient source, either the coil or magnets move relative to each other, voltage will be induced across the ends of the coil. In the simplest configuration, either the magnet of the coil is suspended close to each other. The spring is attached to frame. When vibration occurs either magnet or coil moves relative to the frame of EMEH. Mechanically it is a mass spring system which is also vulnerable to mechanical damping due to air resistance, material stiffness and relative inertias. However electrical damping also occurs as a result of current passing through the inductor. Further information related to EMEHs will be discussed in detail in Chapter 2 of this reports. Detailed literature review supported with schematic and models has been discussed.

The advantage of EMEH over other discussed MEHs is that it doesn't require separate voltage source like electrostatic harvesters nor it is subjected to creep and fatigue like in piezoelectric energy harvesters. EMEHs provides higher energy densities. They are less vulnerable to damages that's results due to sudden spike in vibration pattern or excessive loadings. Due to parallel spring configuration no mechanical stops should be needed. Mechanical stops in MEMS fabrication is a critical issue [20].

1.2.2 Advantages and Disadvantages

1.2.2.1 Advantages

- a. Can generate voltage < 1 volt even on very smaller size.
- b. Can perform over longer period of time than conventional batteries
- c. Can be manufactured using standard MEMS fabrication technology
- d. Can perform on very low to very high frequencies

1.2.2.2 Disadvantages

- a. No standardization so far in this technology
- b. No specific measure of parameters
- c. Very few commercially available energy harvesters

1.2.3 Application of Micro Energy Harvesters

- a. Wireless sensor networks
- b. Bio Medical Implants
- c. Condition and Health monitoring of Machinery
- d. Human Powered Devices
- e. Wearable Gadgets
- f. Health Monitoring in Humans
- g. Structural Monitoring

1.3 Motivation

Micro Electro Mechanical Systems (MEMS) is the emerging technology of this age. Even in its dawn, MEMS find its path in almost every field of technology, from medical sciences to robotics to energy generation systems. It has been three decades, since the initiation of this technology, still a lot of vacuum is prevailing to fill by the researchers. MEMS is the future of every technology and it's just matter of years that MEMS can be seen in every walk of life.

One of the remarkable application of MEMS are micro level energy harvesters. These harvesters provide very low power but can easily be used to power up sensors, condition monitoring systems

and wireless electronics. They have found to be a very good replacement of conventional batteries and can be installed on remote locations. They harvest energy from ambient energy sources from environment thus require no maintenance or replacement over longer time. Observing these benefits, there is a huge need to work on such EMEHs, which will accordingly lead to some reliable solution for such applications.

1.4 Methodology

The course of this research will start with detail literature survey. During the literature review, main focus will be on exploring different mechanism of energy harvesting and types of electromagnetic energy harvesters (EMEHs). After having complete understanding of EMEHs, focus will come down to developing a novel mechanism for Electromagnetic Energy Harvesting and establish state-of-art.

Once the conceptual design will be cleared up, a complete design will be modeled in as a solid model. This task will comprise of creating a solid model for representation of complete design and a simplified model that will be used for Finite Element simulation. Afterward a thorough electromagnetic simulation will be performed in finite element software suite. Derived results will be analyzed and compared with literature. If the results will not up to the mark, few design provisions will be adapted, more preferably changing in magnetic array. In the end final model with complete analysis of novel Electromagnet Energy Harvester will be represented. Figure 1.3 provides comprehensive layout of research activities.



Figure 1.3: Flowchart of Research Activities

Chapter 2. Literature Review

Supply of power through wireless devices are of prime importance and very good replacement of batteries. From the years, significant amount of research has been done in this regard. Energy harvesters are very promising devices in providing wireless technology. Energy harvesters worked on the principle of extracting energy from ambient vibrating source. Energy harvesters finds various application in remote energy consumptions devices like medical implants, wearable computing etc. [21] [21]

Studies indicate that utilization of wireless energy supply system in many applications like medical, industry and military. Usage of such devices minimize the need of external power supplies and batteries. These energy harvesters scavenge energy from the atmosphere ambient sources like temperature, light and vibrations. Vibration is the most promising source in this regard. Because it comes in many forms with abundance like human movement, vehicular motion and seismic vibrations [22].

This study discusses the energy harvesting system based on up-frequency concept. This system generates the electricity through the electromagnetic induction using mechanical movement of the coil in the magnetic field. System consist of top and bottom resonating plates connected with soft springs which damps the ambient vibrations from the source. Operating frequency of the system is between 10-100 Hz ambient frequencies. Simulation results of the system reported the maximum output voltage 6mV and power of 120nW however on micro implementation generates the power of 3.97 μ W at 25 Hz ambient frequency [21]. Figure 2.1 indicates the working model of the said harvester.

Another study has been described by the same group of researchers and they successfully able to improve the previous design using a single cantilever at the frequency of 95Hz with the output voltage of 0.57 mV and power of 0.25 nW with the size of $8.5 \times 7 \times 2.5 \text{ mm2}$ [23].

Another research shows a slightly different approach using same concept of up-conversion of low frequencies. In this system power generation due to the planer motion by using magnet on the low frequency plane and the coils on the high frequency plane. Macro-scale prototype of the said system gives the output of $5.1 \,\mu\text{W}$ and the maximum voltage of 11.1mV. However on conversion

to micro electromechanical systems (MEMS), promised the power of 119 nW and output voltage of 15.2 mV with the size of 10 x 8.5 x 2.5 mm2 [22].



Figure 2.1: 3D model of Micro scale Generator[21]

A novel technique has been introduced by Zorlu et al. using the same concept of up-conversion method for low frequencies. In this study, a magnet has been mounted on a vibrating diaphragm and the coil has been place on a cantilever with hinge that first restrict the motion of cantilever and produce tension the beam and then get released with a jerk and starts vibrating. Figure 2.2 shows the schematic for the discussed EMEHs.



Figure 2.2: Electromagnetic Harvester using magnets mounted on diaphragm. [24] This method converts the low ambient frequency into higher external frequencies with enhances the efficiency of the system. This method can also be adopted for the piezoelectric and electrostatic type of energy harvesters.

However, wearing of the tip of cantilever due to friction and cyclic collision is due which is the short coming in the design. This wearing of the tip is limiting factor of complete life of the

harvester. Resilience and durability of the cantilever tip can be enhanced by proper heat treatment or some coating technique. This energy harvester has been tested to give 88.6 mV and 544.7 μ W with the power density of 184 μ W /cm3. This has been achieved by up-converting the ambient frequency (>10 Hz) to 394 Hz. The volume of this harvester has been reported is 2.96 cm3 [24]. Another a relatively sophisticated vibration based energy harvesting system has been introduced by Arian et al. Proposed system has dual rail, high efficiency system. This system consists of an energy harvester and a PCB mounted with discrete diodes, two capacitors for external output storage and a custom powered ASIC. System converts the kinetic energy into electrical energy through an energy harvester, in which induced voltages from both coils which then rectifies in a bias generator network. Afterwards network converts and step-up the AC into two DC voltages which is saved in the capacitors which is then converted into Dual Rail DC and stored in the external capacitors. All this circuitry is compacted on a single PCB below the harvester module. This system is the size of C-type battery (16cm3). The size comparison can be seen in Figure 2.3. The power density of the system 6.06 micro W/cm3 at 8 Hz vibration. The system delivers the power of 54 μ W [25].



Figure 2.3: EMEH with a series coil harvester [14]

In 2001, a very simple micro level energy harvesting system has been designed. The working principle of this system is based on Faraday's Law. System consist of cantilever with the housing, a magnet has been placed on the beam which provides the magnetic field and a fixed coil. The coil

cuts the magnetic flux and converts mechanical energy into electrical energy. Operating frequency of this system was 320 Hz with the size of 240 mm3 and output power was 1mW. [26].

Another study indicates a simpler yet useful design of micro level energy harvester which can work on frequency range from 50-300 Hz. The magnetic generator uses Neodymium iron Boron (NdFeB) magnets fixed on a steel beam which can vibrate on different frequencies by just adjusting the length of the beam. Design of generator depicts that It can work on resonance frequencies which can be tuned easily to match the frequencies of the environment. This model has been prototyped and tested on a shaker table. However output power and voltages aren't disclosed in the literature [27].



Figure 2.4: Layout of the prototype [27]

Another study proposed two prototype working on the same principle of harvesting energy from ambient vibration source for powering remote accessible sensors and their data transmission. One prototype is consisting of torus magnets and the coil is mounted on cantilever. Coil has been mounted to cut the magnetic field of magnets on vibration of cantilever. However the second prototype has four magnets with the same topology, shown in Figure 2.5. Later one has been tested on top of engine block of 5-years Volkswagen Polo. The average power produced by the generator was 157 μ W [28].



Figure 2.5: Prototype of cantilever based EMEH.

Researcher at University of Southampton, developed a cantilever based micromechanical energy harvester. This energy harvester can function on ambient frequencies as low as 50-60 Hz. The harvester design comprises of four magnets mounted on a cantilever producing constant magnetic flux through the coil as the cantilever vibrates. This energy harvester intent for an air compressor which generates ambient frequency of up to 60 Hz. In this study, first FE simulation has been performed using commercial FEA package. Later on experimental analysis has also been performed. This research paper then summarizes and compares the result of both simulation and experimental results, Experimental results are 25 % lower than the simulation results which mainly due to assembly tolerances in the actual device. Mean output voltage generated b said device is 87mV and power of $17.8 \,\mu\text{W}$ [29].

Another study highlights the usage of low power MEMS based energy harvesters. It has been stated that the current advances in Very Large Scale Integrated Circuits reduces the load of network sensors from mill watts to micro watts. This advancement can remove the batteries as a power source with micro level energy harvesters. These energy harvesters harness the energy from atmosphere using a vibration source. This research developed a theory based design and evaluate it using simulation of electromagnetic harvesters using electroplated micro magnets. Electromagnetic finite element solution has been utilized to calculate the voltage and power which is 55 mV and 70 μ W respectively, also shown in **Error! Reference source not found.** [30].



Figure 2.6: EMEH manufactured by Micro Machined Silicon [31]

Koukhareno et al. proposed a design of electromagnetic energy harvester in which the electromagnets are fixed however the coil is vibrating laterally. This design works on the principle of electromagnetic transduction method. This device operates at very high frequencies ranging from 1.6 - 9.5 kHz generating the peak voltage of 0.7 - 4.15 V and maximum output power of 104 nW [31] [32].

Torah et al. successfully design and implement a micro level energy harvester harnessing energy from ambient power source. This design has been tested and found sufficient to continuously support a radio-frequency (RF) based accelerometer. Power output of the system was 45 μ W with a volume of approximately 150 mm3 [33].

Despite the designing, comparison studies has also been performed by several researchers [34], [35]. In the proposed paper, studies indicate that scaling has the significant effect on the power generation capabilities. Beeby et al. [35] study the effect of scaling from macro to micro level energy harvester power density and found that macro scale energy harvesters exhibit the power density of 2615 nW/mm3. However micro level energy harvester gives the power density of 47 nW/mm3. This study also indicates the difficulty in acquiring ample amount of electromagnetic coupling due to the reduction in size of generator.

Another study indicates the more promising development in the field of energy harnessing devices from environment. Beeby et al. developed a micro electromagnetic harvester with the volume of just 0.15 cm3. In the design phase the, some properties of the harvesters have been optimized to

attain more prominent and practical results. This design employ four laser sintered NdFeB magnets mounted on an etched cantilever using cyanoacrylate as bonding material. The circuit has also been coated with zinc which is used to couple the magnetic flux between top and bottom magnet. This harvester has capability to produce 46 μ W utilizing the vibration frequency of 52 Hz. The power density of this generator is 307 μ W/m3 [36].

Despite the last decade, in recent years, advancements in the MEMS technology give rise to next generation of vibration based energy harvesters. Tao et al. developed a novel mechanism, an adjustable lumped mass coil system in which the harvester can be tuned between frequencies using only 2 DOF configuration. That mechanism gets fabricated using micro machining on a silicon wafer using DRIE technique (double-sided deep reactive-ion etching).



Figure 2.7: Two DOF Electromagnetic Harvester [37]

Usage of this fabrication technique results in very light weight energy harvester i.e. 0.12g. In this proposed research, configuration is in such a way that a permanent magnet is concentrically fixed in the center of a coil on a supporting plate. However the coil has been vibration around the magnet with both connectors are on either side of the coil (top and bottom side), as shown in Figure 2.7. Detailed finite element analysis and electromagnetic analysis has also been done. The power output of the said device is 0.96 nW with the power density of $2.30 \times 10-7 \text{ W/ cm3 g2 [37]}$.

Above discuss power outputs are seemingly more promising than the other related researches in recent year. In 2015, a research indicates the power output of $5.5 \times 10-10$ W using the frequency of 400 Hz with normalized power density of $6.1 \times 10-11$ W /cm3 g2 [38]. Few other research shows the power output of $5.5 \times 10-9$ W, $4.9 \times 10-9$ W, $6.6 \times 10-9$ W with the frequency of 840, 190 and 113 Hz respectively [23], [39], [40].

Yamaguchi et al. presents a novel electromagnetic vibration based energy harvester. Design optimization tools has also been employed to improve the performance of the harvester. In the proposed design the same NdFeB magnets fabricated on Si-Corrugated structure. For improving the harvester power generation rigorous FE analysis has been performed on the trench pitch. Parameters like space ratio and trench pitch has been optimized. On experimentation, stated design produce voltage of 2.65 mV and power of 7.60 nW [41].

In another presented design, a novel completely MEMS based harvester was fabricated in which magnets were sputtered in the silicon substrate trenches and the coils were fabricated on separate substrate and both substrates were then bonded. The coils connected in series were employed in this design. This harvester resulted in output power of 0.12nW but this device had an issue of oxidation of micro magnets [40]. So an improved version of the device was fabricated and the magnets were covered by a polyimide material to prevent the oxidation and the output power improved to 0.760nW [42].



Figure 2.8: MEMS fabricated EMEH [42]

S.	Year	REFERENCE	Frequency Hz	Dimensions	Base	Voltage	Attached	Power	Coil Turns (T)
No.					Acc., g		Load		& Resistance
1.	1995	Williams & Yates [44]	70, 330	5x5x1mm ³				1uW, 0.1mW	
2.	1997	Shearwood & Yates [45]	4.4kHz				39Ω	0.3uW	
3.	1998	Rajeevan & Anantha [46]	94			180mV-p	10 Ω	400 uW	
4.	2001	Ching, Wong et al. [47]	60 - 100	1cm ³		2 - 4.4V p-p	1000Ω	200-830uW rms	
5.	2001	El-Hami, Beeby et al. [26]	320	240mm ³			0.28 Ω	1mW/0.53mW-R1	0.28 Ω
				10.9x3x5.2					
6.	2001	Williams, Yates et al. [48]	0.1-10Mhz	6mm = Lappx		Few mV	39 Ω	0.43 uW	13T, 31 Ω
			4 MHz-p	0.7mm=Tappx					
7.	2004	Beeby et al. [49]	100 Hz device			0.38, 0.76,			71T, 77 ohm (F)
		(2 Coil types, F & Bulk)	6.4, 12.6, 9.5k			0.57Vnl			309T, 109 ohm
			(B)						
8.	2004	Jones, Beeby et al. [50]	A = 322 Hz	$A = 0.84 \text{ cm}^3$			Α = 0.6 Ω	A= 37-180uW	
				B= 3.15cm ³		B=250mV		B= 3.9mW-p,	
								157uW-Avg	
9.	2005	Beeby et al. [51]	9.8, 7.1, 4.7	100mm ³	1g	9, 6.5, 4.3V	20.4, 14.8,	0.5uW/2,	600T
			kHz				9.8	1.4,0.9mW	
10.	2006	Beeby et al. [52]	350 Hz (fig-9)		0.3g	16.9mV	100 Ω	2.85uW	600T
11.	2006	Xingping, Yi-Kuen [53]	40 Hz					35 mW	
12.	2006	Koukharenko, Beeby et al.	1.6kHz, 9.5kHz		1g	0.7V, 4.15V	2 kΩ	0.11mW, 3.35mW	600T, 112 ohm
		[31]							
13.	2006	Kulkarni, Beeby et al. [30]	7.4kHz	30mm ³	0.4g	A= 55mV	A=7.75Ω	A= 70uW	A= 60T/ B=
						B=950mV	B=7.5 kΩ	B= 85uW	250T
									Optimised= 20T
14.	2006	Saha, Beeby et al. [54]	A= 13.11Hz		0.07g	Graph	Graph	Graph	850T/18Ω
			B= 84Hz		7.95g				300T/3.65Ω
15.	2006	Torah, Beeby et al. [29]	56.6Hz		60mg	M1: 52	9 Mohm	1.7uW	600T/100Ω
						mVrms		17.8uW	
						M2: 120			
						mVrms			
16.	2006	Serre et al. [55]	360 Hz	11x11x2.127m				45nW/ 0.2uW	910Ω
				m ³					
17.	2007	Beeby et al. [56]	52 Hz	0.15 cm^3	0.06g	428 mVrms	4 kohm	46uW	2300T
18.	2007	Kulkarni, Beeby et al. [57]	8.08kH &	6mm x 5.7mm	0.39g&1g		52.7 ohm	A=148, B=23nW	A=600T /
			9.83kHz						B=65T
19.	2007	Torah, Beeby et al. [33]	43 – 109 Hz	150mm ³	0.03-0.07g	450mVrms	4 kohm	45uWrms	2300T/1.6 kΩ
20.	2007	Sari, Balkan, Kulah [58]	3.3-3.6 kHz	14x 12.5x8mm ³		50 (E) /	250 ohm	0.5uW	580 Ω
						20mV			
21.	2008	Hohlfeld, Vullers, Boeck	5 Hz	1cm ²		500 mV		100 uW	
		[59]							
22.	2008	Kulkarni, Beeby [60]	A= 808 Hz	A=106 mm ³	0.39g		B=52.7	A= 148nW	Α=600Τ/112 Ω
			B= 9837 Hz	$B = 106 \text{mm}^3$	1g		ohm	B=23nW	B=65T/55 Ω
			C= 60 Hz	$C = 150 \text{mm}^{3}$	0.9g		C=110 ohm	C= 584nW	C=2 times of
									coil B

23.	2008	Kulah, Najafi [61]	1-100Hz (due to	4mm ²		76mV	300 ohm	3.97 uW @ 25 Hz	
			multi-						
			cantilevers)						
24.	2008	Saha et al. [62]	8 Hz	12.7 cm ³	0.03g		7.3 kohm	14.55uW-2.6mW	1000T/800 Ω
25.	2008	Soliman et al. [63]	90-100HZ	0.299cm ³	0.1g	49.3mV	2.7 ohm		22T/ 1.2 Ω
26.	2008	Torah, Beeby et al. [64]	52 Hz	150mm ³	0.06g	931mVrms	15 kohm	58uW	2800T/2.3 kΩ
27.	2008	Zhang et al. [65]	20 Hz	22x16x2.5mm ³		1.36V		3.1uW	66,133T/147,26
									3 Ω
28.	2008	Serre et al. [66]	120 Hz	15x15mm		¬200mV	100 ohm	¬55uW	100 Ω
				memb.					
29.	2009	Ayala, Beeby [67]							
30.	2009	Bouendeu et al. [68]	98 Hz	1.45cm ³	1g	0.27, 0.25V	40 ohm	356 µW	2 * 10.7 Ω
31.	2009	Brian Dick et al. [69]	43.8 Hz	27cm ³	2.3mg (fig-4)	0.9Vrms (fig-		0.55mW (fig 6)	2090T/225 Ω
						4)			
32.	2009	Fondevilla, Serre et al. [70]	344 Hz	1.35cm ³		100mV (fig-	105 ohm	50uW	52T, 100 Ω
						5)			
33.	2009	Hadas, Ondrusek, Singule	17 Hz		0.1g - 0.5g	4Vrms (fig-5)		3mW - 26mW	
		[71]							
34.	2009	Hatipoglu, Urey [72]	46 Hz	30x11.7mm ³	15g	200mV	100 ohm	0.4mW	
35.	2009	Mack, Wallrabe [73]			6.12g	12.5mV	30Ω / 25T		
							each core		
36.	2009	Marioli et al. [74]	102 Hz		1g	183.2mV	76 ohm	290uW (Best)	27.1 Ω
37.	2009	Wang et al. [75]	A=94.5 Hz	0.13cm ³	A=0.5g	42.6mVp-p		A=0.7µW	Α=300Τ/163 Ω
			B=55 Hz		B=1.52g	B=18mV		B=0.61µW	B=60T (2layers)
				2					/ 33Ω
38.	2009	Wang, Arnold [76]	530 Hz	14.3mm ³	1g	13.2Vrms		23pW (Best one)	Resistivity=1.89
39.	2009	P Wang et al. [77]	280 Hz	9x7x5mm ³	1.02g	142.5mV	81 Ω or	17.2μW	2 coils*(30T,
- 10							1MΩ?		40.4Ω)
40.	2010	Z Wang et al. [78]	25 Hz	L=9mm,		2.32 Vp-p	200 Ω		600T
	2010			D=20mm	0.55		101.0		1007
41.	2010	J. Park, Bang, Y. Park [79]	54 Hz	1x1x0.6cm ³	0.57g	68.2mV	18.1 Ω	115.1µW	1831,
42.	2010	B. Yang, C. Lee [80]	40-80Hz (Type	1.5x1.5x1.2cm ³	1.9g	9mV	50 ohm	0.4 uW	12Layers*10T
10	2010		II)	05505 3		12.5 1		20*0.25	200
43.	2010	Sari, Balkan, Kulah [81]	50-200Hz	8.5x/x2.5mm ³		13.5mV		20*0.25nW	20Cantilevers*6
4.4	Ť 2010	T 1 '1 IZ 1 1 A1' [00]	N (1141)	10.05.25 3		(total)		110 W	1/15022
44.	2010	Turkyiimaz, Kulan, Ali [82]	Magnet: 114Hz	10x8.5x2.5mm ³		15.2mV		119nw	921
15	2010	Zhu Daahy at al [92]			0.062	0137		61 6. W 156 6. W	60 5 0T
43. 16	2010	Emra Tan Taral [94]	120 Hz	0.25cm ³	0.00g	0.1 V		01.0μ w - 130.0μ w	10501
40.	2011	Colobay Kim N-1-5 [97]	120 HZ	0.23cm ⁻	1.0	/.2mvp-p	220.0	1.45μ W	12007/2200
4/.	2011	Garrit [42]	10HZ - 05HZ	2.12cm ²	1g	10.4	167.90	105p/15.0avg μ W	12001/ 22002
48.	2011	C. Cepnik [43]	301 HZ	0.46 cm ³	0.01g	10.4mV	107.802	0.62 μ w	137.322
	T								

49.	2011	Q. Yuan, X. Sun et al. [86]	64 Hz	5x5x1mm ³		7.5μV	2 *1.3Ω			1
	*									
50.	2011	X. Xing, X. Sun et al. [87]	42 Hz	29.3cm ³	5g	2.52V	2.5 Ω		20.84mW/cm ³	
51.	2011	E. Bouendeu et al. [88]	132 Hz	8 cm ³	1g	Single coil= 1.1V Series= 4.49V Parallel= 1.07V	1200 Ω 6 kΩ (series) 500Ω (parallel)	1 coil: 108μW 2 coils:404μW (series) & 422μW (parallel)	2100Τ/500 Ω	
52.	2011	Zorlu, E.T.Topal, Kulah [24]	2.96 cm ³			88.6 mV	544.7μW			
53.	2011	Dayal, Dwari, Parsa [89]	108 Hz		Close loop: 2.5g (fig-18)	Both Open & Close Loop= 3.3V	1 kΩ	Open loop:10.8mW Close loop:19.5mW	L	2: (1 (s
54.	2011 *	C. Cepnik, U. Wallrabe [90]		19.8x15.1x3mm 3	1.01g	1.56mVrms	24.9 mΩ	12 μWavg	9.2mΩ	-
55.	2011	Bouendeu et al. [91]	78.43 Hz	20x20mm ² (appx)	1g	0.8(load voltage)	1200 Ω	315.37 µW	675.1Ωinternal	
56.	2011	Foisal, Lee, Chung [92]	9 Hz	14x48mm ²		3.16V_@ 9Hz 4.53V @ 14Hz	97 Ω	1.18mW	1500T/96.502 Ω	
57.	2011	Jiang, Masaoka et al. [40]	115 Hz	0.1cm ³	1.19g	2 mV			1.2kΩ	
58.	2011	J.C Park, J.Y Park [93] (PDMS Springs used)	53 Hz	1x1x0.6cm ³	0.4g	201mVp-p	100 Ω	115.1 μW	183T/ 13.5Ω- Dc	
59.	2011	Galchev et al. [94]	1 - 40 Hz	43(i) / 63(ext.) cm ³	0.01-0.10			0.46-0.72 μW		
60.	2011	Q. Yuan, S. Sun et al. [86]	64 Hz	5x5x1 mm ³		7.5µV				
61.	2011	E. Sardini, M. Serpelloni [95]	FR4: 102 Hz Latex: 41 Hz	8.6cm ³	1g	FR4:183.2m V Latex:378mV	76 Ω	FR4: 290 μW Latex: 153 uW	Inductor:280T/ 22.51Ω, L=24μH/ C=9.5pF	
62.	2011	Zhang et al. [96]	350 Hz	14x13x0.4mm ³		0.27mV	3.90 kΩ	2pW	2 Coils*20T/ 4.77kΩ	
63.	2011	W Pang, L. Wei, C. Lufeng [97]	242 Hz	12x11x? mm ³	0.5g	28mV		0.55 μW	26T	
64.	2011	F. Khan, Sassani, Stoeber [98]	371 Hz (Best)	12x12x7 mm3	13.5g		7.5Ω (best o/p)	max: 23.56µW(pg- 40)	21T/ 7.5Ω	
65.	2012	S.E Joe, M.S Kim, Y.J Kim [100]	8 Hz		Near to 0.5g			430 µW		
66.	2012 *	P. Wang, X. Dai et al. [101]	280.9 Hz	9x7x5mm ³	0.8g	157.5mV	1 ΜΩ	21.7úW	2 Coils in series (Ω not given)	
67.	2012	K. Tao, G. Ding, P. Wang, Z Yang, Y. Wang [102]	365 Hz	20mm ³	1g	20.9 µVp-p (open circuit)			30T	

68.	2012	V. Challa. David Arnold	5 kHz		1g,2g,3g,4g	3.7-16µV	1.5 Ω	9.3, 37.5, 96,	0.8μΩ
		[103]						171pW	
69.	2012	H. Liu, B. W. Soon et al.	1285Hz (Best	10x8x0.45mm ³	1g	3.5-3.6mV	1.8 kΩ	0.016 µW	3
		[104]	mode)						Coils*17T/640,
									000,50022
70.	2012	J. Chen, D. Chen et al. [105]	235 Hz (Best		172mVp-p	90Ω	10µW	90Ω	
			mode)						
71.	2012	Lee, Raman, Chung [106]	12 Hz		0.2g	2.32Vrms	1.8 kΩ	490 µW	2000T
72.	2012	Miki, Fujita et. al. [107]	94.9 Hz		46 coils/ 5T		512 Ω	760 pW	46 Coils/5T
73.	2012	S. Roundy et. al. [108]		37x37x3 mm ³		3Vpeak(fig-8)	20 Ω	11 mJ	
74.	2013	Tanaka, Fujita et. al. [109]	595 Hz			31.3mV		0.722 uW	
75.	2013	K. Ashraf et. al. [110]	10.3 Hz	27.38 cm ³	50mg-1g	7.92V	3 kΩ	20.9mW(5.02mW-	
								avg)	
76.	2013	Z. Li, M.D Han, H. X	48 Hz	2.72mm ³		0.98mv	45 Ω	0.015 μW	8T
		Zhang [14]							
77.	2013	D.H Bang, J.Y Park [111]	A= 36Hz,	0.6mm ³	A&B=0.3g	A&B=107&7		A=29.02µW,	A= 99Ω &
			B=63Hz			3mV		B=23µW	Β=55Ω
78.	2013	H. Liu et. al. [112]	840,1070,1490	0.035cm ³	1g	Vmax=0.003	626 Ω	5.5nW, 0.5nW,	626 Ω
			Hz			V		4.1nW	
79.	2013	H. Lie et. al. [113]	82 Hz	0.032 cm^3	3g	Vmax=0.74m		1.8nW	
						V			
80.	2013	V. Pashaei, M. Bahrami	75 Hz	0.64 cm^3		67.4mV		46.6μW	36T, 24.36 Ω
		[23]							
81.	2013	Q. Zhang, E. Kim [114]	290 Hz	20x5x0.9mm ³		30mV	10.8 Ω	2.6µW	3*20T, 3*3.6Ω
82.	2014	M. Han et. al. [115]	64 Hz	5x5x0.53mm ³	1g	7.5 μV		0.3975nW	1.9 Ω
83.	2014	S. Chen, J. Wu, S. Liu [116]	211Hz, 274Hz	7.1x7x0.04mm ³	1.42g, 1.23g	1&1.03mV		7.66nW, 7.43nW	27 Ω
						(oc)			
84.	2014	M. Han et. al. [117]	26.87 Hz	10x10x10mm ³	0.5g	3.82mV		0.75µW max	2.5 Ω
85.	2014	H. Liu et. al. [118]	62.9 to 383.7	6 x 6 mm ²	0.2g to 1g	0.067 to			
			Hz			0.339mV			
86.	2014	T. Shirai et. al. [119]		38x31x4 mm ³		3.7V			
87.	2014	K. Yamaguchi et al [120]	400 Hz	13.5x14x1.5				3.12 μW	
				mm ³					
88.	2014	Q. Zhang, E.S Kim [121]	400 Hz	51x11x0.4mm ³	6.4g	1.02mV	22 Ω	0.55nW	20Τ, 55Ω
89.	2015	M. Bendame et. al. [122]	12Hz max – (fig		0.6g	0.15V (fig-		12mW	60T (1.75m),
			26)			26)			3.6Ω
90.	2016	K. Tao et. al. [37]	326 Hz, 391 Hz	290mm ³	0.12g	3.6mV,		0.96nW	57T, 12.7kΩ
						6.5mV			
91.	2016	Q. Zhang et. al. [123]	A=250Hz	$A = 0.2 \text{ cm}^3$	A=1.5g		A=0.7Ω	A=14.3µW	Α= 4*7Τ, 0.7Ω
			B=75Hz	B=5.3 cm ³	B=5g		Β=190Ω	B=1.04mW	

NOTE: Base Acceleration, g can be computed by using formula: $g = A * \omega^2$

Table 2.1: Literature Review for Previous work done in the field of micro level harvesters.

Chapter 3. Single Frequency Energy Harvester

As most of the harvesters found in literature work at a single frequency, so initially a harvester design was analyzed at a single frequency of 50 Hz. This chapter discusses the design and results of harvester operating at a single frequency.

3.1 Proposed Harvester Design (mention dimensions in image)

Energy Harvester initially designed for operating frequency of 50Hz comprises of sputtered NdFeB magnets having thickness of 20 μ m and nickel substrate on which magnets are sputtered. Remanent flux density of NdFeB magnets is 1.4T. The magnets are arranged in halbach pattern, so that the maximum flux is concentrated in a single direction (discussed in next section). Springs are added adjacent to the nickel base supporting the magnets and these springs are fixed and subjected to the movement by the external vibrations. These magnets also serve as an inertial mass.



Figure 3.1: Harvester design for the single frequency

3.2 Operating Principle

The proposed harvester works on the principle of Faraday's law that when a magnet moves relative to coil or vice versa, it induces flux in the coils resulting in the voltage output. The movement is caused by the vibrations from the external source.

Mathematically it can be expressed as:

Induced Voltage = EMF = -N *
$$\frac{d\phi}{dt}$$
 (3.1)

Where,

N = No of turns

 $\emptyset = BA = Magnetic flux$

B = Magnetic field

A = Coils Area

The negative sign is due to the Lenz's law caused by eddy currents. The symbols used in the figure 3.1 shows the direction of polarity in accordance to the Halbach pattern.

3.3 Halbach Array

In simple arrangement of magnets, magnetic field are active on both sides of the magnet which has certain drawbacks as well as inefficient utilization of the magnetic flux. On contrary, Halbach array provides efficient and strengthen magnetic flux on active side of the magnet and other side remain quiet with very low or zero magnetic flux. Difference between simple magnetic array and Halbach array is shown in figure below.



Figure 3.2: (a) Simple Magnetic Array. (b) Halbach Array [125]
The flux lines in a simple magnet travel from north to south and so they spread in all directions. Also the maximum flux is concentrated at the edges of the magnet and is less concentrated in the center portion of the magnets. Different orientations are found in literature and figure 3.2 (a) is one of them in which magnets are placed such that the maximum flux is concentrated in between the two pairs of magnets. Other reported orientations of magnets can be seen in [69][43] [123][42][14].

This approach has been utilized in the proposed design successfully. In later sections simulations results shows supremacy of halbach arrangement over simple arrangement. In the proposed design 16 planer magnets were used with main and transit magnets alternatively. Figure below indicates the implementation of halbach array in stepwise manner.



Figure 3.3: Halbach Array in Proposed Micro Energy Harvester

Figure 3.3 shows that how main and transit magnets cancel out the polarities and form the halbach array. It can also be seen that the maximum strength of magnetic flux is on the active side [125]. One more advantage of this arrangement is that if this device is used in electrical circuits, interference of electrical and magnetic field can be avoided.

3.4 Mathematical Modelling

The initial harvester model is designed to work for the vibrational frequency of 50 Hz which can be found in the tires of vehicles and also in the electric machinery.

The computed mass of the magnets and nickel base is 0.000013kg (Appendix A) Using,

$$\omega_{\rm n} = \sqrt{\frac{\rm K}{\rm m}} \tag{3.2}$$

Spring Constant is computed to be, $K_1 = 1.283[N/m]$ at the required frequency of 50 Hz. And for the design of springs, using the following equation,

$$K = \frac{2^* E^* t^* w^3}{L^3}$$
(3.3)

The larger length of the spring comes out to be 2.75[mm]. The smaller length of 46[µm] is optimized such that the natural frequency of 50 Hz is obtained. The thickness and width of 20[µm] and [15µm] are standardized according to the METAL MUMPs process.

Frequency response at multiple smaller lengths of spring (stiffness) can be seen in the table:

S.	Dimension of Smaller	Frequency
No.	length (µm)	(Hz)
1	20	55.155
2	40	51.21
3	46	50.53
4	50	50.89
5	100	48.84

Table 3.1: Optimization for 50 Hz frequency

Initially the damping for the single mass design was computed by using the relations of damping [27] [44] by considering the shear area but that resulted in high amplitudes of excitation which is not desirable. So Rayleigh damping was used to compute the next results. Details can be found in the chapter 4 and Appendix B.

3.5 Modal Analysis of Single Frequency Energy Harvester

In modal analysis, the dynamic behavior of structure under vibrational excitation is studied and mainly the natural frequencies that the device exhibit are observed. In the case of single frequency harvester, the first natural frequency is of our interest and in considered for the further analysis. Also the collective and single axis displacement can be observed. The proposed harvester is designed to move in \pm y axis and the maximum displacement can be observed in this axis. Mainly these modes of vibrations help in suggesting the feasibility of the system when encountered with external vibration source.

During the modal analysis, we subjected Comsol Multiphysics for six frequencies. The damped modal analysis was carried out as can be seen in the figures because when damping is included, the software automatically returns the damped eigen values with real part depicting the natural frequency and the imaginary part showing the damping.

Below mentioned are the six natural frequencies of the said design, we consider the first natural frequency, remaining five are for the visual purpose and showing that they are not desirable due to the non-planer movement.



Eigenfrequency=48.929+1.6139i Surface: Total displacement (µm)

Figure 3.4: First Natural Frequency: (Planar movement)



Figure 3.5: Second Natural Frequency





Figure 3.6 : Third Natural Frequency



Eigenfrequency=115.72+1.6284i Surface: Total displacement (µm)

Figure 3.7: Fourth Natural Frequency

×10⁹ 2 1.5 1 0.5

Eigenfrequency=960.21+3.1093i Surface: Total displacement (µm)

Figure 3.8: Fifth Natural Frequency



Figure 3.9: Sixth Natural Frequency

3.6 Harmonic Analysis of Single Frequency Energy Harvester

The purpose of harmonic study (also called frequency domain study) is to observe the behavior of the system under harmonic force. As the system is linear and the applied force is harmonic so the response of the system is also harmonic.



Figure 3.10: Single Mass: Harmonic Analysis

In the mentioned case for the single mass, the harmonic response shows a single peak showing the maximum displacement of the mass suspended with the springs at the first natural frequency of the system. The input force applied is volumetric with the value of 0.128g for this case. It also shows that as the damping is increased, the system gets slower. And at resonance and no damping, amplitude goes to infinity and there is no convergence at all for the simulation. Hence the plot shows that with damping, system can work at resonance or very near to it.

3.7 Electrical Design

This is layered mechanism, with each layer performing its role with relation to the above layer. The coils are placed underneath the magnets.

3.7.1 Coils

Coils are one of the important component of the harvester. In this design, voltage is induced in the coil by motion of magnets caused due to the vibration from ambient source and the output voltage is taken through the terminal pads of the coil. The coils are planer, micro fabricated with two layers. The coil layers are connected with each other through 4 Via holes each of $2.5E^{-3}$ mm² to maintain conductance and the Via holes are then filled by tungsten. The distance between the coils and magnets is kept 30µm. The standards adopted for coils are of MIMOS. Each coil faces two magnets i.e. a single coil covers an area of 0.625mm x 8mm and the distance is kept symmetric from both ends. This orientation allows the maximum possible flux to interact with the coils. And all the coils are connected in series to achieve a good value of voltage. The thickness of each metal layer is 0.624um and 612µm respectively and a dielectric is used in between two layers so that both the layers are insulated without the interruption of flux lines. The inter turn gap between coil turns is kept 3µm and the width of each turn is also 3µm.



Figure 3.11: Schematic of Coils in Micro Energy Harvester 41

The calculated length of total coil wire is 6.668m with the resistance of $98K\Omega$. Design calculations related to coils are presented in Appendix A.

3.7.2 Circuitry

The following circuit is presented for our harvester model and it has been built up by using the spice model circuitry in Comsol Multiphysics. A low pass filter was also used to avoid the low frequency components. Nodes are defined in the image underneath.



Figure 3.12: Circuit Diagram

Resistance of Coils	R _{coils}	98 kΩ
Inductance of Coils	L _{coils}	0.00353 H
Resistance for L.P Filter	R _{fil} .	98 kΩ
Capacitance for L.P Filter	C _{fil.}	1.624E-8 F
Resistance of Load	R _{load}	98 kΩ
Capacitance of Load	Cload	0.0028703 F

Table 3.1: Circuit Parameters

Where,

$$C_{\rm fil} = \frac{1}{R_{\rm f}^* \omega_{\rm c}} \& \omega_{\rm c} = 2^* \omega_{\rm n}$$
(3.4)

$$C_{\text{load}} = \frac{1}{LC} * \omega_n^2 \& \omega_n = 2 * \pi * f \text{ [rad]}$$
(3.5)

3.8 Output Result for Single Frequency Energy Harvester

Simulation results indicate the following output results for the single frequency energy harvester.



Figure 3.13: Induced and Load voltage of single mass at 48.92 Hz.

As can be seen from above results, a maximum voltage of 22mV is induced, indicated in blue lines represented by V_{out} in the circuitry. As we know that the voltage drops when load is attached and it is represented by the green lines. In most of practical applications where continuous monitoring is not desirable like in tire pressure monitoring system, a capacitor is used for storage of charges and that output is delivered to the external sensor/s after the capacitor is full. And a power of 0.43nW can be extracted (figure 3.14) at a resistive load of 98k Ω , as it can be found in literature that the maximum power can be extracted when the load resistance is equal to the internal resistance [112].

However, the internal resistance can be decreased by using manufacturing standards of other foundry and using coils of gold which will result in an improved output.



Figure 3.14: Induced voltage, Load voltage, Current drawn & Power

The above measurements are made at the first damped modal frequency of 48.92 Hz and its corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.

However, as we know that vibrations are not always of a single frequency, there may be some rise or down in the frequency of vibration source. Also possibly there are fabrication imperfections that rise during the fabrication process like springs which are intended to be of width 15 microns, may be their width is not of exact 15 microns, so it will directly influence the frequency of the device and it possibly will not work at the designed single frequency. So there is need for such a device which can work on a wide range of frequencies, although the peak power may effect but average power will always be greater.

Chapter 4. Dual Mass Energy Harvester with wideband Energy Harvesting

This chapter includes the electromechincal design of the harvester. A novel approach has been utilized to carry out the design of the harvester.

4.1 Proposed Harvester Design

The proposed Micro Energy Harvester also works on the principle of Faraday's Law of electromagnetic induction. The main addition in this device is that an additional mass is added to it such that it harvests energy on a broad range of frequencies.

There are broadly two layers with in the harvester. Figure below shows the mechanical CAD model of said MEH.



Figure 4.1: CAD model of proposed Micro Energy Harvester

Below figure shows the exploded view of harvester model in which the layers can be seen clearly. All of the shown parts will be discussed in the later sections in detail.



Figure 4.2: Exploded view of Micro Electromagnetic Harvester

4.1.1 Parts of Dual Mass Micro Electromagnetic Harvester

4.1.1.1 Substrate:

Silicon substrate grown as the base of the harvester. It's a first layer in the harvester on which the aluminum coil has grown.

4.1.1.2 Coils and Inner Mass:

The coil design and inner magnet design (also inner mass) is kept same and their details have already been discussed in section 3.7.1

4.1.1.3 Outer Mass:

Dual mass system approach has been used while designing the proposed harvester. In principal, this design acts as absorber spring-mass configuration that can be connected to system subjected to unwanted oscillations. As a result, a resonance frequency will be generated between the

unwanted oscillations and spring-mass system [124]. This configuration helps the system to work on the wide band of frequencies.

4.2 Working Principle:

A comprehensible version of Dual mass system is shown in figure below with governing equations. It works such that the Mass M1(outer mass) acts as the drive mass and the Mass M2 (inner mass) acts as the absorber mass. Spring K1 is attached to the substrate and M2 is connected to M1 through Spring K2.

The inner mass, magnets behave as the absorber mass and are responsible for the voltage induction, as discussed in previous sections and the outer mass of nickel acts as the drive mass and is responsible for the wide bandwidth of the electromagnetic energy harvester.

$$m_1 x''_1 + c_1 x'_1 + k_1 x_1 + k_2 (x_1 - x_2) = F$$
(4.1)

$$m_2 x''_2 + c_2 x'_2 + k_2 (x_2 - x_1) = 0$$
(4.2)



Figure 4.3: Dual Absorber Spring-Mass System [124]

Outer mass or inertial mass is attached to inner mass. Outer mass provides the inertia during the vibrations which makes this harvester to perform on multi bands of vibrations. This outer mass is attached to the casing with springs to provide oscillation. Calculations for outer mass and springs are given in Appendix A.

The amplitude and frequency phase response of the dual mass is shown below:



Figure 4.4: (a) Amplitude and (b) phase frequency response of dual mass design [124] The frequencies f_{ol} and f_{oh} are the resonance frequencies and f_{22} is the anti-resonance frequency. More the frequencies f_{ol} and f_{oh} are distant from eachother, more is the bandwidth of the frequencies.

$$\frac{\omega}{\omega_{22}} = \frac{f}{f_{22}} = \sqrt{\frac{1}{2}} \left\{ \frac{1}{\gamma^2} + (1+\mu) \pm \sqrt{\left[\frac{1}{\gamma^2} + (1+\mu)\right]^2 - \frac{4}{\gamma^2}} \right\}^{\frac{1}{2}}$$
(4.3)

The constants μ and γ control the separation between resonant frequencies and the gain respectively. Lesser the μ , less is the frequency separation between two peaks. The following values of μ and γ are used for our design.

$$\gamma = \frac{\omega_{\text{inner}_mass}}{\omega_{\text{outer}_mass}} = \frac{f_{\text{inner}_mass}}{f_{\text{outer}_mass}} = 0.8$$
(4.4)

$$\mu = \frac{m_{\text{inner}_mass}}{m_{\text{outer}_mass}} = 0.5 \tag{4.5}$$

Figure 4.4 (a) and (b) shows separation between the two frequencies and out of phase and in phase response of the frequencies.

4.3 Optimization of g and damping values

Since the damping value calculated by the formulae mentioned in [27] [44] was not feasible giving high values of excitation amplitude, which is totally not desirable for us as it will cause the collision between inner and outer mass. So a procedure was adopted i.e. Design of Experiments Technique (DOE) which uses the central composite design to find the optimum values of g and damping.



Figure 4.5: Optimization of g and damping value by DOE Technique

S.	g	ς	Amplitude	Amplitude	Amplitude at Centre of 1 st
No.	(m/s ²)	Damping	of 1 st Peak (µm)	of 2 nd Peak (µm)	& 2^{nd} frequency (µm)
1	0.255	0.0314	91.6614	35.2075	29.227
2	1.104	0.08403	91.661	35.29	29.228
3	0.6029	0.1815	251.616	92.309	70.3948
4	0.0253	0.018	65.5638	8.3667	3.1347
5	0.7664	0.0622	882.22	214.412	94.09
6	0.128	0.02407	158.06	70.187	15.178

Table 4.1: Different displacement values during optimization of g and damping

4.4 Manufacturing Sequence of Micro Energy Harvester

Manufacturing of the Micro Energy harvester starts with high-resistive silicon wafer which act as base which is manufactured using photolithography and deep reactive ion itching (DRIE). Right after that coils will be manufactured with aluminum using electroplating on the slicing base. This layer will be completed by developing bond using KMPR photoresist, patterned with photolithography.

For the second layer it starts with Silicon frame and high-aspect-ratio outer springs which are manufacture by DRIE and photolithography. Afterwards outer mass of nickel will develop using bulk micromachining. Again high-aspect-ratio springs will be developed using reactive ion etching. These spring will then be connected with inner mass. Inner mass comprises of high-resistive silicon wafer layer and magnet array. Silicon wafer layer will be manufactured using DRIE, with magnet molds created in it. NdFeB magnet array will develop using DC magnetron sputtering. Due to sputtering high performance magnetic properties will be achieved [40].

In the end, both layers can be bonded using already created KMPR photoresist by heating both layers under certain pressure. A manufacturing layout of proposed Micro Energy Harvester is presented in schema below.



Figure 4.6: Manufacturing layout Scheme with cross-section

The cross-sections in figure 4.6 are taken from the locations as mentioned in the below mentioned figure.



Figure 4.7: Dual Mass design representing the X-section at different lengths

Chapter 5. Finite Element Simulation Approach

CAD models are primarily generated for the manufacturing and visualization purposes that is the reason modelers keep it detailed, even minute hinge is model. This detailed model is the "dream" of CAD modeler. Contrary it become "night mare" for the CAE personnel.

Finite Element Analysis performs calculations on nodes of an object. It has been observed from the studies and experience that number of complex geometry features are not required to perform an analysis [125]. Inclusion of those features not only increases the calculation cost but also produces some unnecessary results. So it has been widely advised by experts to simplify geometry as much as possible. But during this process extreme care should be taken not to omit the feature which is directly essential for the analysis.

There are some additional features in any model which are quite irrelevant for simulation and don't play any role in results of analysis but increase the computation time and cost.

5.1 Geometry Formation:

Creating geometry in COMSOL is all about simplification of the CAD geometry. Complete geometry for the proposed system is shown in Figure 5.1 below.



Figure 5.1: Model for Finite Element Analysis

Following are the steps which were taken in order to create a simplified model with the necessary features for the analysis.

5.2 Steps for modeling the system for FEA

 It starts with the creation of block for base plate with the dimensions 8mm x 5mm x 0.02mm. Nickel was assigned to it as material property.



Figure 5.2: Base Plate of Nickel

 After base, an array of 16 magnets was modeled, each magnet of dimension 8mm x 0.3125mm x 0.02mm. Magnetic properties assigned to this array has been discussed in details in later sections.



Figure 5.3: Array of 16 magnets

3. Springs were formed on the four sides of the nickel base plate & magnets resulting in the following geometry.



Figure 5.4: Spring Design for Inner Mass

Above geometry is also termed as inner mass in proposed harvester. Dimensions of springs connected to inner mass are as follows:

Smaller length: 0.046mm x 0.015mm x 0.02mm

Larger length: 2.758mm x 0.015mm x 0.02mm

The dimension and visibility of inner mass helps to create the geometry of outer mass.

4. Top and bottom blocks are buildup for outer mass each of dimension 12.172mm x 3mm x 0.02mm.



Figure 5.5: Step 1 of 2 of outer mass modeling

5. Then further two blocks are buildup, one each for top and bottom mass of dimensions 11.336mm x 3.221mm x 0.02mm, resulting in the following geometry.



Figure 5.6: Step 2 of 2 of outer mass modeling

- 6. Finally the springs are formed connected to outer mass as follows:

Figure 5.7: Springs of outer mass

Dimensions of springs connected to outer mass are as follows:

Smaller length: 0.0234mm x 0.01mm x 0.02mm

Larger length: 1.9002mm x 0.01mm x 0.02mm

Union option in Booleans and Partition is used to make sure that the faces of smaller and large lengths of springs attached to inner and outer mass are properly connected to each other. Both the inner and outer masses are designed to move in a single axis, y-axis by using the dual mass design approach discussed before and its calculations can be found in Appendix A.

5.3 Assigning Analysis Parameters

It is obvious that to perform analysis simplification the geometry to the maximum extent to avoid the computation complexities especially due to the deformation physics used in the simulations which is used to study the changes in physics when the geometry (represented by mesh in this case) changes due to imposed change in geometry externally (vibrations in this case).

For the analysis, it is known that inner mass (magnets) is responsible for the voltage induction, so for simplification and ease, the geometry was minimized by removing the springs and outer mass, as can be seen in the Figure 5.8.





The shaded portion in the Figure 5.8 are magnets.

5.3.1 Physics and Boundary Conditions:

Two physics are used in the analysis, Magnetic Fields, No Current and Deformed Geometry. The following boundary conditions and selections are used for the Magnetic Analysis:

- 1. Initially the magnets are centered w.r.t the origin.
- 2. The air block is defined enclosing the magnets. Magnets can be seen inside.



Figure 5.9: Boundary Condition for magnets

3. The coils area for four coils is defined (shaded portion) such that one coil is facing two magnets and the coils are symmetric from both ends.



Figure 5.10: Coil definition in analysis

- 3. The air block is partitioned from the center such that magnets are in one portion and the coils are in the other portion. It is done because magnets are moving and are given an input sinusoidal in deformed geometry whereas the coils are kept stationary, so they are in the other portion.
- 4. In the end to save the computation time, geometry is further reduced by using the Partition Objects option and finally we simulate the half portion of our geometry.



Figure 5.11: Centered Half Section for Computing optimization

- 5. The Form Assembly option is used instead of form Union at the end of geometry formation. Due to this Comsol deals geometry as a group of the geometry objects instead of treating all parts as a single geometry and pairs are formed.
- 6. The Zero Magnetic Scalar Potential is defined in magnetic field, no current (mfnc) physics and any edge is selected at the outer face of air box. It provides the boundary condition that depict zero magnetic potential on the boundary.
- 7. Magnets are assigned polarities in the Magnetic Flux Conservation option as follows:

•	Magnetic Field			•	Magnetic Field		
Cor	stitutive relation:			Cons	stitutive relation:		
R	emanent flux density		•	Remanent flux density 🔹			
B =	$= \mu_0 \mu_r \mathbf{H} + \mathbf{B}_r$			B =	$\mu_0 \mu_r \mathbf{H} + \mathbf{B}_r$		
Relative permeability: μ_r From material		μ_r From material					
Ren	nanent flux density:			Rem	anent flux density:		
Br	0 Br 0	x y z	т	Br	0 0 -Br	x y z	т
	Magnetic Field			*	Magnetic Field		
Cor	stitutive relation:			Con	stitutive relation:		
Remanent flux density 🔹			Remanent flux density 👻				
B = Rela	$= \mu_0 \mu_r \mathbf{H} + \mathbf{B}_r$ tive permeability:			B = Rela	$= \mu_0 \mu_r \mathbf{H} + \mathbf{B}_r$ tive permeability:		
μ_r From material \bullet		μ_r From material \bullet					
Ren	nanent flux density:			Rem	nanent flux density:		
	0	x			0	x	
			3,05,5	D	0	1.3	
Br	-Br	У	ा	Br	0	У	т

Figure 5.12: (a) (b) (c) (d) Definition of four different magnet polarities

- 8. The Figure 5.12 shows the assignment of polarities of magnets in four different directions for the Halbach Array. Br is the Remanent flux density and its value is defined as 1.4T in parameters. The negative sign reverses the polarity direction of magnet and finally the same pattern is used for the remaining twelve magnets.
- In the deformed geometry physics, A*sin(2*pi*f*t) function is used as a vibrational input to our magnets in y direction.
- 10. The default value of 1.0E-6 of Relative repair tolerance is used.

5.3.2 Meshing:

Meshing is the discrete representation of geometry. There are four types of element types, TET, bricks, prisms, and pyramids. Free Tetrahedral mesh is used for the meshing of our geometry. Since magnets are of much interest, so they are meshed finely. Below mentioned are the meshes of our geometries.



Figure 5.13: Tetrahedral meshing of the harvester model



Figure 5.14: Meshing of bounded region (air)

Chapter 6. Results and Discussion

In this section, first the results of magnetic analysis have been discussed. In the magnetic analysis flux density for Halbach magnetic array has been calculated. In further sections in this chapter modal, voltage and power output have been calculated.

6.1 Flux Density

Flux density is the density of magnetic lines in a specific region. Using Halbach array, flux density of the system has been enhanced which is shown in Figure 6.1. Contour plots indicate that due to halbach array flux density is maximum at the active side of the magnet. However it is very low on the quite side of the magnet.



Figure 6.1: Isometric view of magnetic array and flux density.



z component (T) Arrow Surface: Magnetic flux density (Spatial) Arrow Surface:

Figure 6.2: 2D – Z Component of Halbach Magnetic Flux Density Lines



Figure 6.3: Flux density along coils at 20µm distance

It is clear from the Figure 6.3 that flux density is maximum at -2 mm to 2 mm in the y-coordinate which is 0.21 T



Figure 6.4: Flux density along coils at 30µm distance: (Case of Thesis)



MAGNETIC FLUX DENSITY ALONG COILS, z COMPONENT (T). 100 µm from magnets



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However, increasing the distance between coils and magnets will result in weak flux density. This phenomena is shown in Figure 6.4 and Figure 6.5 at the distance of $30 \,\mu\text{m}$ and $100 \,\mu\text{m}$ respectively. It is important to note that Figure 6.4 is the case adopted in this thesis.

6.2 Modal Analysis

In modal analysis of double mass, natural frequency of the system has been carried out and first two modes have been taken into account which results in two peaks as discussed in previous section. These modes of vibrations help in suggesting the feasibility of the system when encountered with external vibration source.

6.2.1 Dual Mass

For dual mass, first two natural frequencies are of our interest. The results are displayed in the images underneath.



Figure 6.6: First Natural Frequency: (movement along y-axis)



Figure 6.6: Second Natural Frequency

6.3 Harmonic Analysis

Here we will observe the harmonic response of the dual mass design. In both cases, the force applied is volumetric with the value of 0.128 g. The Rayleigh Damping constants α and β are 0.02407.



Figure 6.8: Dual Mass: Harmonic Analysis

Maximum amplitude can be seen at 40.7 Hz and 87.6 Hz.

6.4 Transient Analysis

Transient analysis indicate the displacement behavior of the system in any instance. In this study transient analysis has been performed on dual mass system only.



Figure 6.9: Transient Analysis of Dual Mass System

It can be seen that harvester become stable after 0.2 seconds and deliver output at constant rate.

6.5 Output Voltage

Voltage and Power results for the double mass are presented in the following chapter. All power calculations are done at a load resistance of $98k\Omega$ which is equal to that of the coil resistance, as can be found from literature that the maximum power can be extracted when internal resistance is same to that of the external resistance. [112]

In dual mass system investigation, five different frequencies have been taken into consideration while calculating the output voltage and power.

Case-I: 40.7 Hz (First peak of harmonic Analysis)

As can be seen from below results, a maximum voltage of 18mV is induced and 0.3nW power can be extracted. The above measurements are made at the frequency of 40.7Hz and its corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.



Figure 6.7: Dual Mass Induced and Load voltage of single mass at 40.7 Hz



Figure 6.11: All parameters of dual mass system at 40.7 Hz.

Case-II: 87.6 Hz (Second peak of harmonic Analysis)

As can be seen from below results, a maximum voltage of 14.5mV is induced and 0.2nW power can be extracted. The above measurements are made at the frequency of 87.6Hz and its corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.



Figure 6.12: Dual Mass Induced and Load voltage of single mass at 87.6 Hz



Figure 6.13: All parameters of dual mass system at 87.6 Hz.

Case-III: 100 Hz

As can be seen from above results, a maximum voltage of 1.5mV is induced and 4.84pW power can be extracted. The above measurements are made at the frequency of 100Hz and its

corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.



Figure 6.14: Dual Mass Induced and Load voltage of single mass at 100 Hz



Figure 6.15: All parameters of dual mass system at 100 Hz.
Case-IV: 60 Hz

As can be seen from above results, a maximum voltage of 2.32mV is induced and 7.5pW power can be extracted. The above measurements are made at the frequency of 60Hz and its corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.



Figure 6.16: Dual Mass Induced and Load voltage of single mass at 60 Hz



Figure 6.17: All parameters of dual mass system at 60 Hz.

Case-V: 20 Hz

As can be seen from above results, a maximum voltage of 0.67mV is induced and 0.215nW power can be extracted. The above measurements are made at the frequency of 87.6Hz and its corresponding excitation amplitude is obtained from the interpolated function extracted from the harmonic analysis.



Figure 6.18: Dual Mass Induced and Load voltage of single mass at 20 Hz



Figure 6.19: All parameters of dual mass system at 20 Hz.

6.6 Conclusion

As can be seen from the above results that using the dual mass approach, we can successfully harvest energy at multiple frequencies. Also the methodology adopted to widen the bandwidth is successfully implemented by using the dual mass approach, although this method reduces the peak power output but increases the average power output across its bandwidth. This can clearly be seen from the results. Also higher the frequency, more is the rate of change of flux and hence it is also an important parameter for the harvester design but in dual mass case, excitation amplitude is a dominating factor.

The output power is maximum at 40.7 Hz as expected due to the higher excitation amplitude. Then at the frequency of 80.7 Hz, we get the second maximum output where there is the second maximum value of excitation amplitude. If we consider the frequency of 60Hz between the two peaks obtained from harmonic analysis, we get the 3rd maximum output as this frequency lies in between the bandwidth area of harvester. And comparing the results to the frequencies other than the band between two peaks, 20Hz and 100 Hz, it can be seen that the output power is less as compared to peak frequencies due to the less gain values as compared to that of the 60Hz.

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Appendix A

INNER MASS COMPUTATIONS

• Inner Mass:

Using,

Mass= Density x Volume & Volume=Length x Width x Thickness Volume of Magnets = 8[mm] x 5[mm] x $0.02[mm] = 0.8[mm^3] = 8^{E-10} [m^3]$ Mass = 0.00000588[kg]Volume of Ni Plate = 8[mm] x 5[mm] x $0.02[mm] = 0.8[mm^3] = 8^{E-10} [m^3]$ Mass = 0.0000712[kg]Total Mass = 0.000013[kg]

• Inner Spring Constant:

Using,

$$\omega_n = \sqrt{\frac{K}{m}}$$

 $K_{inner} = 1.283[N/m] @ f = 50[Hz]$

• Inner Spring Larger Length:

Using,

$$K = \frac{2^* E^* t^* w^3}{L^3}$$

L = 2.75[mm]

OUTER MASS COMPUTATIONS

• For frequency of Using dual mass design approach:

 $\frac{\omega_{inner_mass}}{\omega_{outer_mass}} = \frac{f_{inner_mass}}{f_{outer_mass}} = 0.8$

 $f_{outer} = 62.5[Hz] @ f_{inner} = 50[Hz] \& \omega = 2*\pi*f$

• Outer Spring Constant:

Using,

$$\omega_n = \sqrt{\frac{K}{m}}$$

 $K_{outer} = 4.0095[N/m]$

• Outer Spring Larger Length:

Using:

$$K = \frac{2*E*t*w^3}{L^3}$$

L = 1.9002[m]

NOTE-1: In Comsol, width of outer spring was kept 10µm instead of 15µm to get in plane motion

• For outer mass computation, using dual mass design approach:

 $\mu = \frac{m_{inner}}{m_{outer}} = 0.5$

 $m_{outer} = 0.00026[kg] @ m_{inner} = 0.000013[kg]$

• Mass distribution in equal proportion in four parts:

Using,

Density= $\frac{\text{mass}}{\text{volume}}$

Equivalent volume of Nickel Mass $(0.00026 \text{ kg}) = 2.9213\text{E}^{-9}[\text{m}^3]$

Vertical Length = 2*[0.046mm+0.046mm+0.046mm+0.015mm] (combined length of springs on top and bottom) + 5[mm] (height of magnets) + 6[mm] (width of top and bottom mass = 3mm each kept constant)

= 11.336[mm] = 0.011336[m]

For Left and right mass Dimensions: Using, Volume=Length x Width x Thickness Using values above: Volume = 7.3033E⁻¹⁰[m³], Length = 0.011336[m], Thickness = 0.2E⁵[m]

Width of left horizontal mass = width of right horizontal mass = 0.003221[m] = 3.221[mm]

For Top and Bottom mass Dimensions: Remaining Volume= $\frac{2.9213E^{-9}}{2} = 1.4606 E^{-9}[m^3]$ One side Volume= $\frac{1.4606E^{-9}}{2} = 7.3033E^{-10}[m^3]$ Using, Volume=Length x Width x Thickness Using these values above: Volume = 7.3033E⁻¹⁰[m³], Width = 0.003 (kept constant) [m] & Thickness = 0.2E⁵[m] Length of top horizontal Mass = Length of top vertical mass = 0.00121722[m] = 12.1722[mm] NOTE-2: Smaller lengths were designed to get internal mass frequency of 50[Hz] and outer

mass frequency of 62.5[Hz]

Coil Calculation:

Width = $3[\mu m] = 0.000003[m]$ L = $6.25E^{-4}[m] \ge 8E^{-3}[m]$ • L = $\frac{6.25E^{-4} \ge 8E^{-3}}{3^{-6}} = 0.8333[m]$ for single coil

Total length = $0.8333 \times 8 = 6.668[m]$

- $R = \rho \frac{1}{A} = 98[k\Omega]$ at $A = 6.125E^{-7}[\mu m]$
- No of Turns:

#Matlab code for No of Turns

width = 0.625*10^-3; height = 8*10^-3; area = width * height; wireLengthLimit = 1; wireWidth = 3*10^-5; gap = 3*10^-5;

no_Turn = -1; wireLength = 0;

while (area>0 && wireLength<wireLengthLimit)</pre>

```
no_Turn = no_Turn + 1;
```

```
width = width - (2*gap) - (2*wireWidth);
height = height - (2*gap) - (2*wireWidth);
```

area = width * height;

wireLength = wireLength + (2*width + 2*height - gap);
end
no_Turn

Appendix B

Analytical Modelling

An electromagnetic harvester is considered as the mass-spring system with a proof mass m, which is suspended by the spring of constant K and that spring is held by some external frame. So when the vibrations take place, the mass (magnets) moves and its relative motion w.r.t coil placed nearby induces the voltage. Air acts as a damper medium.

The governing equation for power harvesting is presented by William & Yates [43] of a system can be expressed as:

$$P = \frac{m^* \xi_{\tau} * Y^2 * \left(\frac{\omega}{\omega_n}\right)^3 * \omega^3}{\left[1 - \left(\frac{\omega}{\omega_n}\right)^2\right]^2 + \left[2^* \xi \tau^* \left(\frac{\omega}{\omega_n}\right)\right]^2}$$

When the system works at resonance, we get the maximum output. So it is necessary to design the device with the natural frequency equal to that of the external vibration.

$$P = \frac{m^* \, \xi \tau^* \, Y^{2*} \, \omega_n^3}{4^* \xi^2}$$

Y = Excitation amplitude

 ω_n = Natural Frequency

 ξ_{τ} = Total damping in the system comprising of electrical and mechanical damping, expressed in the next portion.

From above equation, we can see that the Power is proportional to the cube of natural frequency and square of excitation amplitude.

Damping:

Damping is an important parameter while designing a MEMS device. It is one of the basic parameter that controls the selectiveness of device especially if we want to focus any particular application. Since in the current design, a wideband of frequencies is required, so a high damping factor is necessary. [45] Its effect is considered in the harmonic analysis of design, mentioned in the chapter of Results.

Other widely used methods to increase the bandwidth include the use of multiple array of different magnetic masses of different dimensions [23] and magnetic field tuning methodology [25]. But

the increase of damping method works best, especially when there are plentiful vibrations of multiple frequencies.

Damping is of two types that mainly take place, Parasitic and Electrical damping. Parasitic damping is caused mainly caused by the surrounding medium, air in our case and few of its effect is due to the acceleration. Whereas Electrical Damping is caused due to the internal resistance of the coils.

An important point during designing of a harvester is the matching of resonant frequency of harvester to that of the source of vibration and the maximum possible deflection is desirable to get the maximum rate of change of flux. And thus by varying the electrical impedance of load, we can control the damping and thus adjusting the frequency selectivity of the harvester.

Mechanical damping can be computed by using the formula [126]:

$$C = \frac{\mu^* A}{h}$$

And using this value of C in the below equation gives us the mechanical damping:

$$\xi_{\rm m} = \frac{C}{2^*m^*\omega_{\rm n}}$$

Where ω can also be used in terms of spring constant, K by using:

$$\omega_n = \sqrt{\frac{K}{m}}$$

Electrical Damping is calculated as follows:

$$\xi_{\varepsilon} = \frac{(B^*l)^2}{2^*m^*\omega_n^*R}$$

Where,

C = Damping co-efficient

 $\mu = effective viscosity of fluid$

h = Gap between two surfaces where damping is to be considered

m = mass of device (Calculated in Appendix)

 $\omega_n = 2 \pi^* f$ = Natural Frequency in Radians

B = Magnetic Field Strength

l = length of coil

B = Magnetic field strength

R = Resistance of the coil (Calculated in Appendix)

Rayleigh Damping:

Also termed as Alpha-Beta damping is a practical damping model and can be estimated from experiments. Also unlike other models, it can be used in both time dependent and frequency dependent (harmonic) analysis. It is used when there are different frequency responses. Since dual mass design approach is used, so in the simulations, Rayleigh damping is taken into account.

Option for Rayleigh damping is found in the structural Mechanics module. In this damping two damping coefficients are mentioned.

Rayleigh Damping = $C = \alpha M + \beta K$

 α = mass proportional Rayleigh damping coefficient

 β = stiffness proportional Rayleigh damping coefficient

M = system structural mass matrix

K = system structural stiffness matrix

NOTE: $\alpha \& \beta$ should be positive

Rayleigh damping coefficients can also be expressed in terms of f₁ and f₂ as:

$$\alpha = 4 * \pi * f_1 * f_2 \frac{\xi_1 f_2 - \xi_2 f_1}{f_2^2 - f_1^2}$$
$$\beta = \frac{\xi_2 f_2 - \xi_1 f_1}{\pi (f_2^2 - f_1^2)}$$

 $\alpha \& \beta = 0.0240757$ is used in the simulations and this value was found by using the <u>Design of</u> <u>Experiment Technique</u> (DOE), which uses Response Surface Methodology algorithm.

Appendix C

Expressions used in COMSOL Multiphysics®:

Expression	Description	
Nc*intCoilA(mfnc.Bz)	Flux in Coil A (replace A by	
	B,C,D for other coils)	
fluxA + fluxB + fluxC + fluxD	Single layer flux	
-d(totalFlux2, TIME)	Total induced voltage	
comp1.cir.v_4*comp1.cir.C2_i	Power absorbed by load	

• Material Properties used in Comsol for Modal Studies:

Property/Material	Unit	Nickel	NdFeB
Density	Kg/m ³	8900	7350
Young's Modulus	Pa	200E ⁹	10E ¹¹
Poisson's Ratio	1	0.31	0.24

• Material Properties used for Magnetic Study:

Resistivity of Aluminum used = $2.65[\mu\Omega-cm]$

Relative permeability of Air = 1